Design and Optimization of a Novel Bored Biplanar Permanent-Magnet Assembly for Hybrid Magnetic Resonance Imaging Systems

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We present a novel design for a biplanar permanent-magnet assembly suitable for use in hybrid magnetic resonance imaging (MRI) systems. The key feature of our design is a large cylindrical hole that is longitudinally bored through the entire magnet assembly. The presence of the bore permits the potential inclusion of additional peripheral devices within or near the magnet structure that may benefit from being placed along, or oriented parallel to, the main magnetic field. In particular, the magnet assembly can be considered for use in an integrated system consisting of a 6 MV medical linear accelerator (linac) coupled to an MRI system for state-of-the-art real-time image-guided adaptive radiotherapy. We use magnetic field calculations based on the finite-element method (FEM) to quantify the detrimental effects of the bore on the field homogeneity in the imaging volume for pole-piece designs found commonly in industry. We then demonstrate that shape optimization of the pole pieces yields novel pole designs that lead to suitable levels of field homogeneity. We examined the resultant magnetic field within the bore for the optimized design and found that it has maximum field homogeneity.

Index Terms—Geometrical parameterization, image-guided radiotherapy, magnetic field homogeneity, magnetic resonance imaging (MRI), nonaxisymmetric, permanent magnet, shape optimization.

I. INTRODUCTION

MAGNETIC RESONANCE IMAGING (MRI) is a powerful noninvasive biomedical technique capable of producing high-quality images with exceptional soft-tissue contrast. Recent advancements in MRI technology coupled with the ability to perform functional and spectroscopic imaging have contributed to the rapidly growing popularity of this imaging modality.

MRI systems generally employ permanent, resistive, or superconducting magnets to produce a strong homogeneous static magnetic field $B_0$ that uniformly polarizes the nuclear magnetic spins of the subject examined. Superconducting magnets offer the greatest field strengths and highest image quality, thus optimizing the design of these systems has been of great interest [1]–[12]. Resistive magnets produce weaker field strengths than superconducting systems, but they are less expensive and exhibit smaller fringe fields. As a result, optimal design methods for compact resistive magnets have also been pursued [13]–[16]. Unfortunately, resistive and superconducting magnets require dedicated cooling and power systems, and are relatively expensive to build and operate. Consequently, permanent-magnet systems are emerging as a viable alternative in low-field MRI applications, due to their relatively low weight, size, and cost. Resistive and superconducting magnet designs are typically based on cylindrical or split-coil windings; therefore, these systems are restricted in terms of their possible geometrical configurations [17]. In comparison, a large variety of open and compact permanent-magnet geometries can be found [18]–[26]. Advancements in permanent-magnet design optimization [18], [27]–[37], and post-manufacturing passive shimming [38]–[41], have opened the door for magnet designers to pursue new and unique permanent-magnet configurations tailored to specific applications.

In this work, a design is proposed for a novel biplanar permanent-magnet structure with a large cylindrical hole longitudinally bored through the entire assembly. The structure geometry is based on a four-column design as first presented in [27]. The existence of the bore permits the inclusion of additional therapeutic or diagnostic devices within or proximate to the magnet structure that might benefit from being directed along, or oriented parallel to, the main magnetic field of the MRI system. In particular, this magnet structure is considered a candidate for use in an integrated medical linear accelerator (linac) and MRI system capable of performing real-time image-guided adaptive radiotherapy.

The use of a biplanar permanent magnet in the design of an integrated linac-MRI system was first proposed by our group in [42] and [43]. In the initial design of this system, a 6 MV linac was coupled to the open end of a 0.2 T permanent magnet assembly such that the linac and X-ray beam were oriented perpendicular to the main magnetic field. It was later reported that this transverse arrangement of the X-ray beam with respect to the main magnetic field leads to the production of unfavorable hot and cold spots at tissue-air interfaces in the resultant treatment plans at high field strengths [44]. Recently, an alternative arrangement was proposed by our group in which the linac and X-ray beam were oriented parallel to the main magnetic field [45]. It has been shown that this parallel arrangement is advantageous in that it leads to a reduction of the aforementioned hot and cold spots observed with the transverse arrangement [45], [46]. A schematic diagram illustrating both arrangements is shown in Fig. 1.

In order for a biplanar permanent magnet system to permit a parallel arrangement, a large hole needs to be vacated from the magnet assembly in order to provide an unobstructed beam path for the linac. It is expected that the presence of such a hole would affect the characteristics of the magnetic field in the imaging volume and thus impact the quality of images produced by this system. Consequently, an investigation into the viability of a
bored biplanar permanent magnet for MRI is necessary so that the desired parallel arrangement can be pursued. Furthermore, the proper functioning of a linac device may be affected by the presence of external magnetic fields through interactions with the accelerating electrons in the waveguide interior. Detailed characterization of the magnetic field produced by the magnet system at potential linac locations is required so that a complete simulation of the electron dynamics in the waveguide can be performed.

In this paper, the finite-element method (FEM) is employed to quantify the effects of the large bore on the magnetic field homogeneity in the imaging volume. It is verified that flat or single-ring annular shimmed (SRAS) pole pieces, as found commonly in industry, are inadequate in yielding suitable levels of field homogeneity for medical imaging [27], [30], [32], [35]. Accordingly, the optimization scheme previously presented in [37] is adapted to yield a novel optimized double-ring annular shimmed (DRAS) pole piece, as well as optimized axisymmetric and non-axisymmetric grid parameterized (AGP and NAGP) pole pieces. The method employed is similar in philosophy to the target-field approach presented in [47] for the design of circular biplanar gradient and shim coils, which relies on analytical field expressions valid in the absence of magnetic materials. In addition, the axisymmetric designs obtained are similar to the strictly axisymmetric solutions presented in [27], [35] for permanent-magnet assemblies without a large bore.

Magnetic field inhomogeneities corresponding to the various pole piece designs are compared and the necessity of the optimized DRAS, AGP, and NAGP designs is demonstrated. The transverse and longitudinal magnetic fields produced by the magnet assembly with the NAGP optimized design are examined within and proximate the large bore, providing insight towards the necessary details required in designing an integrated linac-MRI system with a parallel arrangement.

II. Finite-Element Method Problem Formulation
The magnetic field calculations required in this work are performed with the 3-D FEM software package COMSOL MULTIPHYSICS (version 3.4, Comsol AB, Stockholm, Sweden). Details of the FEM analysis have been provided in [37] and are reproduced here. The FEM problem formulation begins with Maxwell’s equations in a current free region:

\[ \nabla \times \mathbf{H} = 0 \]  
(1-a)

\[ \nabla \cdot \mathbf{B} = 0 \]  
(1-b)

where \( \mathbf{H} \) is the magnetic field vector and \( \mathbf{B} \) is the magnetic flux density. Due to the absence of currents, \( \mathbf{H} \) can be expressed in terms of the gradient of a scalar potential \( V_m \):

\[ \mathbf{H} = -\nabla V_m. \]  
(2)

In a region with magnetic permeability \( \mu \) and magnetization \( \mathbf{M} \), the fields \( \mathbf{H} \) and \( \mathbf{B} \) are related by

\[ \mathbf{B} = \mu (\mathbf{H} + \mathbf{M}). \]  
(3)

Combining (1-a), (1-b), (2), and (3) yields

\[ \nabla \cdot (\mu \nabla V_m - \mu \mathbf{M}) = 0, \]  
(4)

The magnetostatics application mode in the ac/dc module of COMSOL MULTIPHYSICS is employed to numerically solve (4) for \( V_m \) at discrete points within the model geometry, subject to appropriate boundary conditions. The fields \( \mathbf{H} \) and \( \mathbf{B} \) can then be calculated directly by (2) and (3), respectively.

Convergence is achieved in solving the nonlinear problem (4) when the specified relative tolerance of \( 10^{-6} \) exceeds the relative error estimate, which is computed as a weighted Euclidean norm of the estimated nodal error values in \( V_m \) [48]. Tetrahedral quadratic Lagrange elements are used to partition the model geometry in order to obtain an accurate piecewise linear solution for \( \mathbf{H} \) and \( \mathbf{B} \) without excessive computational burden.

III. A Novel Bored Permanent-Magnet Assembly
The bored four-column biplanar permanent-magnet assembly considered in this work is shown in Fig. 2. This magnet assembly is identical to the reduced magnet structure analyzed in [37] with the exception of a large 20 cm diameter longitudinal hole bored through the center of the entire magnet assembly. Table I provides values for the dimensions labeled in Fig. 2(b), where the minimum pole separation \( g \) is chosen as 70 cm to allow for rotation around a typical human patient.

The FEM calculations take into account the nonlinearity of the AISI 1020 plain carbon steel yoke and column structures, and the Armco magnetic steel pole pieces, through the relative permeability functions shown in Fig. 3 [37], [49], [50]. Additionally, the residual flux density of the Nd-Fe-B permanent magnet poles is chosen as 1.005 T.

In order to compare various magnet assembly designs, the inhomogeneity \( \operatorname{PPM}_V \) of the magnetic field measured over the volume of interest \( V \) is introduced as a value of merit:

\[ \operatorname{PPM}_V = \frac{B_{\text{max}} - B_{\text{min}}}{B_0} \times 10^6 \]  
(5)

where \( B_{\text{max}} \) and \( B_{\text{min}} \) are the maximum and minimum values for the magnetic flux density magnitude within \( V \), respectively, and \( B_0 \) is the magnetic flux density magnitude at isocenter.
TABLE I
BORED PERMANENT-MAGNET ASSEMBLY DIMENSIONS

<table>
<thead>
<tr>
<th>Quantity</th>
<th>Length (mm)</th>
</tr>
</thead>
<tbody>
<tr>
<td>h</td>
<td>2066</td>
</tr>
<tr>
<td>g</td>
<td>700</td>
</tr>
<tr>
<td>w</td>
<td>1578</td>
</tr>
<tr>
<td>l</td>
<td>2600</td>
</tr>
<tr>
<td>d</td>
<td>1398</td>
</tr>
<tr>
<td>r</td>
<td>100</td>
</tr>
</tbody>
</table>

Fig. 3. Relative permeability functions for Armco magnetic steel and AISI 1020 plain carbon steel.

The effect of the large bore is to reduce the field at isocenter $B_0$ as the amount of magnetic material in the assembly that contributes to the net magnetic field at this point is reduced. This reduction is especially significant for the design with the SRAS pole piece; the optimized peripheral ring has greater depth in the bored system, resulting in a further reduction of magnetic material as compared to the standard system without the bore.

The presence of the bore also leads to a dramatic increase in the $\text{PPM}_Y$ inhomogeneity for both flat and SRAS pole piece designs. Based on the assumption that successful post-manufacturing shimming of a magnet assembly requires an inhomogeneity value of less than 300 ppm in the imaging volume of interest, the bored magnet systems examined here are considered unsuitable for medical imaging over both 30 and 40 cm DSVs.

IV. GEOMETRIC PARAMETERIZATION OF THE DESIGN SURFACE

A. Annular Shimmed Parameterization

The open face of the magnetic pole piece of the bored magnet assembly is taken as the design surface. The design variables are selected as the dimensions of the pole piece cross section, as shown for the upper pole piece in Fig. 4. This results in five unique design variables for the DRAS design, and three unique design variables for the SRAS design with the constraint $z_4 = z_5 = 0$.

B. Grid Parameterization

For the grid-based designs, the open face of the magnetic pole piece is parameterized with 112 control points distributed on the polar grid shown in Fig. 5. This parameterization is similar to that used in [37], with the appropriate adjustments in order to account for the presence of the bore. The $z$ coordinates of the control points are taken as the design variables so that the thickness of the pole piece at each of these locations is varied during optimization. The complete pole piece surface is then generated based on a linear spline interpolation between the control points, similar to the Bezier spline approach in [35].

Particular symmetry in the pole piece surface allows for constraints to be imposed on the control points in order to reduce
TABLE II
COMPARISON OF MAGNETIC FIELD PROPERTIES

<table>
<thead>
<tr>
<th>Quantity</th>
<th>Standard Magnet Assembly</th>
<th>Bored Magnet Assembly</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Flat Pole Piece</td>
<td>SRAS Pole Piece</td>
</tr>
<tr>
<td>Iterations</td>
<td>-</td>
<td>11</td>
</tr>
<tr>
<td>$B_n$ (T)</td>
<td>0.202</td>
<td>0.179</td>
</tr>
<tr>
<td>$\Psi(10^{11} \cdot \text{T}^{-1} \cdot \text{m}^{-3})$</td>
<td>1440</td>
<td>1.895</td>
</tr>
<tr>
<td>$PPM_{40cm}$</td>
<td>8734</td>
<td>321</td>
</tr>
<tr>
<td>$PPM_{80cm}$</td>
<td>15232</td>
<td>1030</td>
</tr>
<tr>
<td></td>
<td>Flat Pole Piece</td>
<td>DRAS Pole Piece</td>
</tr>
<tr>
<td>Iterations</td>
<td>-</td>
<td>75</td>
</tr>
<tr>
<td>$B_n$ (T)</td>
<td>0.197</td>
<td>0.157</td>
</tr>
<tr>
<td>$\Psi(10^{11} \cdot \text{T}^{-1} \cdot \text{m}^{-3})$</td>
<td>647.1</td>
<td>19.51</td>
</tr>
<tr>
<td>$PPM_{40cm}$</td>
<td>7130</td>
<td>1355</td>
</tr>
<tr>
<td>$PPM_{80cm}$</td>
<td>24978</td>
<td>2116</td>
</tr>
<tr>
<td></td>
<td>Axisymmetric Optimized Pole Piece</td>
<td>672</td>
</tr>
<tr>
<td></td>
<td>Non-Axisymmetric Optimized Pole Piece</td>
<td>1022</td>
</tr>
</tbody>
</table>

Fig. 4. Design parameterization for optimization of the SRAS and DRAS pole piece designs. Only the upper pole piece is illustrated.

Fig. 5. Control point distribution for the AGP and NAGP pole piece designs, plotted in cartesian coordinates.

V. NONLINEAR OPTIMIZATION ALGORITHM

Details of the iterative optimization process have been presented in [37] and are reproduced here. The major steps of the iterative scheme are summarized in the flow diagram of Fig. 6. The algorithm is scripted in MATLAB (version 7.5, The MathWorks Inc., Natick, MA) and utilizes macros from the COMSOL MULTIPHYSICS scripting language. The optimization can be expressed as

\[
\begin{align*}
\text{Minimize } \Psi &= f(\mathbf{p}) \quad \text{(6-a)} \\
\text{Subject to } p_{\text{min}} &\leq p_i \leq p_{\text{max}}, \text{ for } i = 1, \ldots, N \quad \text{(6-b)}
\end{align*}
\]

where $\Psi$ is the objective function and $\mathbf{p}$ is the design vector. For $N$ unique design variables, the elements of $\mathbf{p}$ are denoted $p_i$ (for $i = 1, \ldots, N$) and are subject to the lower and upper constraints $p_{\text{min}}$ and $p_{\text{max}}$, respectively.

The objective function minimized during optimization is defined as

\[
\Psi = \int_{\Omega} (B(\mathbf{r};\mathbf{p}) - B_0)^2 \, d\mathbf{r}
\]

where $B(\mathbf{r};\mathbf{p})$ is the magnitude of the magnetic flux density at the point $\mathbf{r}$, implicitly due to the magnet design defined by $\mathbf{p}$. Therefore, $\Psi$ is a cumulative measure of the magnetic field inhomogeneity in the volume $\Omega$, which is chosen as a 40 cm
DSV at isocenter. The gradient of the objective function $\nabla \Psi$ with respect to the design variables is defined as

$$\nabla \Psi = \begin{bmatrix} \frac{\partial \Psi}{\partial p_1} & \cdots & \frac{\partial \Psi}{\partial p_N} \end{bmatrix}$$

(8)

for which each of the partial derivatives is approximated using a forward finite difference approximation:

$$\frac{\partial \Psi}{\partial p_i} \simeq \frac{\Psi(p_i + \delta p) - \Psi(p_i)}{\delta p}$$

(9)

with a perturbation size $\delta p$ of 0.1 mm. In order to minimize (7),
the design vector is updated at the $k$th iteration according to the well-known steepest descent optimization method [51]:

$$p^{k+1} = p^k - \alpha^k \nabla \Psi^k$$

(10)

where $\alpha$ is the step size determined by a simple bisection line search algorithm [52].

The iterative optimization process is repeated until convergence in the objective function is achieved.

VI. NUMERICAL OPTIMIZATION RESULTS

An optimized DRAS pole piece design was obtained after 186 iterations, while optimized AGP and NAGP designs were obtained after 672 and 1022 iterations, respectively. The resulting designs are visually compared in Fig. 7 alongside the flat and SRAS designs. Magnetic field properties for the various designs are tabulated and compared in Table II.

The greatest field uniformity is obtained with the NAGP design, which results in the lowest values for the objective function and inhomogeneity over both 30 and 40 cm DSVs. The performance and geometry of the AGP design is very similar to the NAGP design, with both designs now permitting post-manufacturing shimming and medical imaging over a 30 cm DSV. Evidently, the NAGP design does not provide a significant advantage over the AGP design, as the largest inhomogeneities present are those due to the bore and not the nonaxisymmetric configuration of the yoke structure and columns.

The simple and effective DRAS design achieves significantly improved field inhomogeneity values compared to the flat and SRAS designs. Notably, it yields a 30 cm DSV that is suitable for post-manufacturing shimming and medical imaging. Although the inhomogeneity values for this design are 2–3 times greater than the AGP and NAGP designs, the optimization requires significantly less computation time and the resulting design would be simpler to construct.

VII. MAGNETIC FIELD ANALYSIS

The magnetic field produced by the bored magnet assembly with the NAGP optimized pole piece design is analyzed in terms of its longitudinal $B_z$ and transverse $B_\perp$ components within and proximate to the large bore. Longitudinal profiles passing through the bore are shown in Fig. 8 for a selection of positions along the $x$ axis.

The various $B_z$ profiles obtained are relatively similar in their shape and magnitude. A common trait of these profiles is a drop from the value near isocenter to a large negative value within the bore, followed by an increase to a small positive value that asymptotically approaches zero. The region of negative $B_z$ within the bore corresponds to where the magnetic field reverses direction, as can be seen in the magnetic field line plot of Fig. 9.

In contrast, although the $B_\perp$ profiles retain a similar shape at various positions, the peak magnitude of this component changes dramatically. $B_\perp$ is zero on the $z$ axis due to symmetry about the center of the bore, and increases in magnitude with increasing position along the $x$ axis, as the bore edges are approached. This effect is illustrated in Fig. 8(c), which shows the peak magnitude of the transverse component within the bore volume as a function of distance $x$ from the $z$ axis. This field behavior and the shape of the profiles are congruent with Fig. 9, as the magnetic field reversal within the bore necessitates a change in the polarity of the transverse component as well.

As the profiles indicate, the magnetic field within the bore is complex and highly inhomogeneous. In particular, both the longitudinal and transverse components reverse direction in this region, with the total field magnitude peaking at nearly 0.3 T.
In comparison, the region exterior to the bore exhibits a weaker field with less variation and therefore may lend itself as a better candidate for the location of a magnetically sensitive device such as a linac. Although increasing the distance between the linac and isocenter would result in a reduction of the maximum achievable dose rate, this region is less confined and would better accommodate the necessary additional equipment and shielding required for actual linac operation. In any case, the complete three dimensional vector description of the magnetic field produced by the magnet assembly has been obtained and can now be used to simulate the electron dynamics of the linac waveguide in order to precisely evaluate its performance.

VIII. CONCLUSION

The design and optimization of a novel bored biplanar permanent-magnet assembly for MRI has been presented. A unique feature of this system is a large longitudinal hole that has been vacated from the magnet assembly to permit the inclusion of additional devices within or near the magnet structure that may be oriented along the direction of the main magnetic field. It has been demonstrated that the presence of the large bore degrades the magnetic field homogeneity in the imaging volume and therefore optimized pole piece designs are required. A complete description of the magnetic field produced by the magnet assembly has been obtained and will contribute to the further development of an integrated linac-MRI device with a parallel arrangement.

ACKNOWLEDGMENT

This work was supported in part by the Alberta Heritage Foundation for Medical Research and the Natural Sciences and Engineering Research Council of Canada.

REFERENCES


